

A Method for Distorted Pulse Restoration Based on Frequency Domain Analysis and Transformer Model

Authors: Mr. Dejie Chen, Huang, Prof. Hong-Quan, Wu, Mr. Zhong-Lin, Dr. Xingke Ma, Wang, Qiu-Yi, Lan, Zhi-Yong, Nie, Mr. Li, Huang, Prof. Hong-Quan

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Abstract

To address the issue of complex distortion in the output signals of the front-end analog system during nuclear pulse signal detection, this paper proposes a distorted pulse restoration method that combines frequency domain analysis with the Transformer model. This method accurately identifies the relevant parameters of nuclear pulse signals through a combination of frequency domain analysis and sample learning, thereby correcting the distorted pulse signals. Initially, pulse signals are analyzed using frequency domain analysis to ascertain the property of the components within the distorted pulses (such as inertial, proportional, and integral components) and the initial parameters of each component. Subsequently, based on the initial parameter ranges of the identified pulse components and predefined mathematical models, multiple datasets comprising distorted pulse sequences are constructed to train the Transformer model. In simulations, the trained model is validated using simulated pulses. Taking distorted first-order, second-order, and third-order pulse signals as examples, the corresponding model is employed to identify their amplitudes, and the relative errors are computed. The results indicate that, compared to the case without frequency domain analysis, the incorporation of frequency domain analysis reduces the relative errors in amplitude parameter identification by 0.81%, 0.87%, and 0.77%, respectively. Finally, to further verify the feasibility of combining frequency domain analysis with the Transformer model and assess the model's performance, five X-ray measurements are conducted on a copper ore sample. The measured pulse sequences are digitally processed and used as input to the model. Multi-channel pulse amplitude analysis is then performed on the output pulse amplitudes from the model, yielding an X-ray energy spectrum corrected for shadow peaks. The model's performance is evaluated using the metric R, which represents the proportion of shadow peaks corrected by the model. The

results demonstrate that approximately 94.15% of the shadow peaks can be corrected to characteristic peaks. The method proposed in this paper, which combines frequency domain analysis and a transformer model, can effectively restore distorted pulse signals and is of great importance for improving the completeness and resolution of the radiation energy spectrum count.

Full Text

Preamble

A Method for Distorted Pulse Restoration Based on Frequency Domain Analysis and Transformer Model

De-Jie Chen, Hong-Quan Huang*, Zhong-Lin Wu, Xing-Ke Ma, Qiu-Yi Wang, Zhi-Yong Lan, Li Nie

(College of Nuclear Technology and Automation Engineering, Chengdu University of Technology, Dongsanlu, Erxianqiao, Chengdu, 610059, People's Republic of China)

Abstract

To address the issue of complex distortion in the output signals of the front-end analog system during nuclear pulse signal detection, this paper proposes a distorted pulse restoration method that combines frequency domain analysis with the Transformer model. This method accurately identifies the relevant parameters of nuclear pulse signals through a combination of frequency domain analysis and sample learning, thereby correcting the distorted pulse signals. Initially, pulse signals are analyzed using frequency domain analysis to ascertain the property of the components within the distorted pulses (such as inertial, proportional, and integral components) and the initial parameters of each component. Subsequently, based on the initial parameter ranges of the identified pulse components and predefined mathematical models, multiple datasets comprising distorted pulse sequences are constructed to train the Transformer model. In simulations, the trained model is validated using simulated pulses. Taking distorted first-order, second-order, and third-order pulse signals as examples, the corresponding model is employed to identify their amplitudes, and the relative errors are computed. The results indicate that, compared to the case without frequency domain analysis, incorporation of frequency domain analysis reduces the relative errors in amplitude parameter identification by 0.81%, 0.87%, and 0.77%, respectively. Finally, to further verify the feasibility of combining frequency domain analysis with the Transformer model and assess the model's performance, five X-ray measurements are conducted on a copper ore sample. The measured pulse sequences are digitally processed and used as input to the model. Multi-channel pulse amplitude analysis is then performed on the output pulse amplitudes from the model, yielding an X-ray energy spectrum corrected for shadow peaks. The model's performance is evaluated using the metric R, which represents the proportion of shadow peaks corrected by the model. The

results demonstrate that approximately 94.15% of the shadow peaks can be corrected to characteristic peaks. The method proposed in this paper, which combines frequency domain analysis and a transformer model, can effectively restore distorted pulse signals and is of great importance for improving the completeness and resolution of the radiation energy spectrum count.

Keywords: Deep learning, Transformer model, Trapezoidal shaping, Pulse signal restoration

1. Introduction

In the digital processing and energy spectrum analysis of nuclear pulse signals, trapezoidal shaping emerges as a frequently employed signal processing technique, effectively facilitating the identification of pulse amplitudes. However, the ubiquitous issue of pulse distortion in practical applications poses a severe challenge to the accuracy of these amplitudes. In measurement systems equipped with switch-reset preamplifiers, distorted pulses primarily consist of truncated pulses, which occur when the pulse amplitude abruptly drops to zero due to the switching reset, resulting in pulse signals with insufficient effective width. This distortion problem leads to significant amplitude loss after trapezoidal shaping, ultimately giving rise to shadow peaks near the full-energy peak in the energy spectrum, thereby compromising the counting accuracy and resolution of characteristic peaks.

Various repair methods can be adopted to address this issue, including deep learning approaches and repair algorithms. However, when distortion issues arise in complex pulses, the repair difficulty increases substantially. For instance, in certain nuclear radiation detection applications, the complexity of the front-end analog channel, comprising detectors and conditioning circuits, along with the instability of electronic component parameters induced by temporal or environmental changes, results in output pulse signals containing multiple components. These components exhibit diverse characteristics in the frequency domain, including proportional, inertial, integral, differential (encompassing first-order or higher-order), and oscillatory components. In the time domain, they manifest as complex waveform features such as exponential signals, step signals, ramp signals, and oscillatory signals. The diversity of pulse components significantly elevates the complexity of the signals. When these pulse components undergo distortion, if they cannot be effectively identified and processed, the task of restoring the distorted pulses becomes even more arduous, ultimately impacting the restoration effectiveness and leading to reduced counting accuracy and resolution of characteristic peaks.

In the field of nuclear signal processing and energy spectrum analysis, an increasing number of researchers are utilizing new algorithms and technologies to address domain-specific challenges. Zhang et al. investigated the effect of CR-RC digital filter parameters on nuclear energy spectrum measurements and obtained optimal parameters. To achieve high count rates, a new true Gaus-

sian digital shaper for detector pulses and a compensation technology for pulse stacking were proposed. V.T. Jordanov et al. proposed efficient digital pulse shaping algorithms for real-time synthesis of multiple pulse shapes. The effect of parameter drift of trapezoidal shaping in nuclear pulse processing was investigated and a correction method was proposed to enhance system robustness. A bi-exponential signal pulse shaping algorithm was proposed for high count rate environments with improved signal-to-noise ratio and energy resolution. Huang et al. investigated a genetic algorithm-based method for pile-up nuclear pulse parameter estimation.

In addition, deep learning technology has rapidly developed in recent years and demonstrated extensive application potential in the field of nuclear signal energy spectrum analysis. For example, a fast nuclide identification method based on ANN, SVM and other algorithms was proposed to help achieve rapid identification and classification of gamma and X-ray spectra in nuclear applications. Griffiths et al. used a convolutional neural network to screen neutron-gamma pulses, extracting features directly from the raw data with superior performance in identifying potential substructures of the signal. A method using a specific NN called U-net as a pulse shaper was developed, which filters the pulses and returns them unfolded, thus solving the stacking problem and even estimating the pulse amplitude when the detector is saturated. Byoungil et al. proposed a deep learning based method for separating and predicting the true pulse height of a signal for application in spectroscopy with a scintillation detector. Deep learning technology provides various ideas for pulse processing in radiation measurement. Touch et al. applied an artificial neural network for energy spectrum correction and achieved satisfactory results. Tang Lin et al. proposed using the CNN-LSTM model to identify relevant parameters of distorted pulses and simultaneously achieved energy spectrum correction. Holl et al. identified pulse shapes in germanium detectors by combining a neural network for feature extraction with a smaller classification network. Wang et al. applied the Transformer model to identify pulse parameters. Ai et al. generalized uncertainty estimation of nuclear detector signals based on neural networks and integrated learning. Ma et al. used deep learning CNN-LSTM models for estimating trapezoidal overlapping nuclear pulses.

However, for the aforementioned pulse signals with multiple components and complex distortions, a single deep learning model often struggles to fully recognize and effectively process all distortion scenarios. Therefore, this paper proposes an innovative method for restoring distorted pulses, which combines the advantages of frequency domain analysis and the deep learning Transformer model. This method can deeply analyze the complex waveform characteristics of pulse signals, fully recognize and process various distorted components, thereby effectively restoring distorted pulses and improving the accuracy and resolution of nuclear signal energy spectrum analysis. Introducing this technology into the field of nuclear signal energy spectrum analysis is of great significance for enhancing the counting accuracy and resolution of characteristic peaks.

2.1 Frequency Domain Analysis

The frequency domain analysis method is used to fully identify the components contained in the complex pulses output by the analogue front-end system of a nuclear radiation detector, and to obtain the property and initial parameters of each component. The implementation process is as follows:

Step : Randomly select a complete single pulse from the measured pulses, plot the frequency characteristic trend line, and calculate the slope of the trend line as well as the coordinates of the intersection points between trend lines, using the following method: Find the frequency domain expression for the pulse as shown in Eq. (1), where K is a positive constant. Plot the horizontal coordinate ω and $= 0 \ln |H(j\omega)| = \frac{K}{\omega}$ as the vertical coordinate, and plot the trend line K on the curve with straight line. Assume the following form of the pulse s-domain obtained from the measurement pulse:

The individual components included in this are as follows: one inertial component: , component number “1” ; one oscillatory component: , with component number “2” ; one differential component: $+ 1 + 1$, component number “3” . The trend lines are shown in Figure 1 [Figure 1: see original paper], and there are 4 trend lines for this pulse. The intersection of trend line K1 and the vertical coordinate is recorded with serial number “0” , with slope 0. The intersection coordinates of trend line K1 and trend line K2 are recorded with serial number “1” . The intersection coordinates of trend line K2 and trend line K3 are recorded with serial number “2” . The intersection coordinates of trend line K3 and trend line K4 are recorded with serial number “3” .

Step : Construct the slope matrix and the component matrix of the trend line, and establish the correspondence between the slope matrix, the component matrix and the coordinates of the intersection points from Step , implemented as follows: The possible slopes of trendline 2, trendline 3, and trendline 4 form the slope matrix U as follows:

From left to right, the three elements in the same row of matrix U correspond to the slopes of trendline 2, trendline 3 and trendline 4, respectively. Combining the slope characteristics of the three component trend lines and deriving the corresponding component matrices W from the slope matrix U :

where the elements of W are formed by the ordinal numbers of the components. The correspondence between U , W and intersection points: taking the first row of U as an example, correspond to the slopes of trendline K2, trendline K3 and trendline K4, and correspond to the component numbers 1, 2 and 3 of the first row of W (i.e., inertial, oscillatory, differential components), corresponding to intersections respectively. This correspondence is summarized as shown in Eq. (5):

Step : Use the slope of the trendline of the pulse and the coordinates of the intersection points from Step to find the initial parameter values of the function $G(s)$ as follows (a)~(c):

- (a) The inertial component parameters K and λ are solved as follows: In Eq. (6), “j” denotes the sequence number of the element whose value is “1” in the three elements $\text{# } 6$, and it is obtained from $0 = 0 \ln K$.
- (b) The differential component parameter τ is solved as in Eq. (6). In Eq. (8), “l” denotes the sequence number of the element whose value is “3” in the three elements $\text{# } 8$.
- (c) The oscillatory component parameters are solved according to Eq. (9): In Eq. (9), “q” denotes the sequence number of the element whose value is “2” in the three elements.

2.2.1 Pulse Model

As mentioned above, due to the complexity of the front-end analog channel composed of the detector and conditioning circuit, as well as the instability of electronic component parameters in the circuit caused by time or environmental changes, the output pulse signal contains multiple components. These components exhibit characteristics such as proportional, inertial, integral, differential, and oscillatory in the frequency domain, and manifest as complex signals such as exponential signals, step signals, ramp signals, and oscillatory signals in the time domain. As shown in Figure 2 [Figure 2: see original paper], these pulse signals can be obtained by connecting the detector output to different circuits (such as RC integrator circuits, differentiator circuits, etc.). Among them, the first stage is the equivalent circuit at the detector output. Each stage is connected through an ideal voltage amplifier, assuming that its input impedance is infinite, output impedance is zero, frequency band is infinitely wide, and voltage gain is 1. Figure 3 [Figure 3: see original paper] presents the s-domain block diagram of the circuit.

In Figure 3, u_d denotes the detector output, u_1 is the output signal of the first stage RC circuit, and the simplification of the s-domain block diagram yields (s) as shown in Eq. (10):

This paper takes first-order, second-order, and third-order pulses as examples to derive their mathematical models. Firstly, u_1 , u_2 , and u_3 represent the time-domain models of the first-order, second-order, and third-order pulses, respectively. By expanding and simplifying the fraction in Eq. (10), we obtain the output of the third-order RC circuit:

In Eq. (11): Q , and Q is the electric charge. To obtain the third-order pulse time-domain mathematical model u_3 , u_3 . Let $b=0$, and then inverse transform Eq. (11). Discretization of Eq. (12) leads to a third-order pulse digitization model:

In Eq. (13): $u_3 - u_3 - u_3 - u_3 - u_3 = 1 / (u_3 - u_3)$, where $u_3 - u_3 = u_3 / (u_3 - u_3)$ are unequal to each other; A is the pulse amplitude $u_3 - u_3 - u_3$ coefficient of the third order, $u_3 / (u_3 - u_3 - u_3)$. From Eq. (12), when $u_3 = 1$ is the sampling period, the first-order pulse time-domain expression and the digitized model 1

$\alpha = 1$, $\beta = 0$ can be obtained as follows: $\beta = 0$. In Eq. (14): α is the first-order pulse amplitude coefficient, τ is the time constant, and T_s is the sampling period. From Eq. (12), when $\beta = 0$, the second-order pulse time-domain expression $1 = 1$ can be obtained as follows: $\beta = 0$ and the digitized model. In Eq. (16): α is the second-order pulse amplitude coefficient, when $\alpha = 1$ is negative, τ , in this paper for the experimental simulation of take the value of 1, fast time constant, is the sampling period.

2.2.2 Pulse Digital Shaping Model

In the context of digital shaping of nuclear pulses, trapezoidal shaping assumes a significant role, serving as a replacement for complex analog filter shaping circuits. The high signal-to-noise ratio performance of the pulses post-filtering, coupled with the capacity to adjust the pulse width and flat-top width of the output waveform by tuning the shaping parameters, renders trapezoidal shaping a prevalent component in digital measurement systems. The time-domain expression for the trapezoidal pulse is given by Eq. (18):

In the equation given by (18), the symbol A denotes the flat-top amplitude of the trapezoidal pulse, τ signifies the rise time of the trapezoidal pulse, β designates the shaping time of the trapezoidal pulse, $u(t)$ is the ideal step function. Subsequent to performing the z-transform on the pulse, the following result is obtained:

In Eq. (19): γ , and T_s is the sampling period.

(1) Trapezoidal shaping of first-order pulse: The z-domain expression for the pulse of first order is obtained by transforming Eq. (14) by z as follows:

The transfer function for the first-order pulse trapezoidal shaping transform can be obtained by associating Eqs. (19) and (20) as follows: (21). Cross-multiplying both sides of Eq. (21) equation and further simplifying using the inverse transformation of z yields the following recursive formula for first-order pulse trapezoidal shaping:

In Eq. (22): γ is the proportionality coefficient between the trapezoidal amplitude and the first-order pulse amplitude coefficient A , $\gamma=1$, is the sampling period.

(2) Trapezoidal shaping of second-order pulse: Similarly, the z-transformation of Eq. (16) and the simultaneous coupling of Eq. (19) leads to a recursive equation for trapezoidal shaping of second-order pulses as follows:

In Eq. (23): γ is the proportionality coefficient between the trapezoidal amplitude and the second-order pulse amplitude coefficient A , $\gamma = 1$, and T_s is the sampling period.

(3) Trapezoidal shaping of third-order pulse: Similarly, the z-transformation of Eq. (13) and the simultaneous coupling of Eq. (19) leads to a recursive equation for trapezoidal shaping of third-order pulses as follows:

In Eq. (24): $T = T_1 = T_2 = T_3 =$ is the sampling period; $2(T_1 + T_3) + \gamma$ is the proportionality coefficient between the trapezoidal amplitude $3(T_1 + T_2)$ and the third-order pulse $2T_1T_3 + 3T_1T_2$ amplitude coefficients A, $\gamma=1$.

In summary, a digital model for the pulse and a trapezoidal shape expression have been derived, which will be used as an input to the deep learning model.

3.1 Frequency Domain Analysis and Transformer Model Principle Analysis

To address the complex distortion of output signals from the front-end analog system in nuclear pulse signal detection, this study employs a method that combines frequency domain analysis (hereinafter referred to as FDA) with the deep learning Transformer model. The implementation process is as follows: (1) The nuclear pulse signals output by the front-end analog system are analyzed using FDA to determine the property of the components contained in the pulse (such as differentiation, inertia, proportion, integration, and oscillation), and the initial parameter values of each component are calculated. (2) Based on the property of the pulse components and the range of their initial parameter values, hierarchical training is conducted to obtain different Transformer models. (3) The measured distorted pulse signals are converted into digital signals $X(n)$ through a conditioning circuit. (4) The digital signals $X(n)$ undergo digital waveform shaping to yield $Y(n)$. (5) Depending on the range of initial parameter values obtained in (1) and the property of the components contained in the distorted pulse, $Y(n)$ is fed into a specific submodule within the corresponding Transformer model for parameter identification, thereby obtaining more accurate parameters such as pulse amplitude. The flowchart for the restoration of complex distorted nuclear pulse signals is shown in Figure 4 [Figure 4: see original paper].

By integrating frequency domain analysis with a hierarchical Transformer model architecture, complex distorted pulses are processed to achieve the restoration of distorted pulse signals. The specific working principle is as follows: Taking the first-order pulse signal shown in Figure 4 as an example, the process initiates with the application of frequency domain analysis to identify the components within the nuclear pulse, yielding the first-order pulse and its initial parameter value. Subsequently, a digital shaping technique is employed to apply trapezoidal shaping to the distorted first-order pulse, resulting in the trapezoidal pulse. Finally, based on the property of (the trapezoidal shaping outcome of the first-order pulse) and the range of, etc., is routed to the first-layer model (Transformer1) for further processing. However, the specific sub-model within the Transformer1 first layer that is selected depends on the range in which the initial value falls, determined by whether τ , where τ is the time constant used during model training. This approach significantly enhances the accuracy of the model, as it focuses on analyzing pulses with specific characteristics and within a narrow parameter range, thereby reducing the complexity of processing distorted pulses and improving the model's precision.

As mentioned above, during the model training phase, different models are trained based on the property of the pulses and the range of their initial parameter values. Therefore, when constructing the dataset, this paper generates distinct datasets according to the range of the initial value τ_0 of the first-order pulses, which are then used to train different Transformer models. For instance, when training Transformer models for first-order pulses, the relationship between different models and their corresponding datasets is illustrated in Figure 5 [Figure 5: see original paper]: the dataset for the Transformer1 model covers a time constant τ range of [3.2, 5.0] s; whereas, the dataset for the Transformer1' model spans a time constant τ range of [6.0, 10.0] s. During model invocation, if $\tau = 3.8$ s, the Transformer1 model is selected; if $\tau = 7.9$ s, Transformer1' model is chosen; and so forth. This approach narrows down the search range for the pulse parameter τ , enabling the selection of an appropriate Transformer model for analysis, thereby significantly enhancing the accuracy of the model.

3.1.1 Transformer Model Feasibility Analysis

Nuclear pulse signals are digitally processed to convert them into waveforms with specific characteristics. As shown in equation (24), a third-order pulse can be transformed into a trapezoidal pulse, suggesting a complex fitting relationship between the trapezoidal pulse and the shaping parameters A , τ . These parameters can be accurately identified using fitting algorithms, such as deep learning neural network models, to restore distorted pulse signals. (Here, τ , where represent the rise time and flat-top duration of the trapezoidal shaping, respectively; τ denotes the time constant, and A represents the pulse amplitude coefficient).

The Transformer model, with its powerful sequence processing and feature extraction capabilities, can effectively identify the trapezoidal shaping parameters of pulse signals, thereby restoring distorted pulses. Therefore, the core of addressing the distorted pulse problem lies in using the model to identify the trapezoidal shaping parameters of the pulse and using these parameters to replace the parameters of the original distorted pulse, thus achieving pulse restoration. The specific implementation steps include: dataset construction (generating a pulse sequence dataset containing different shaping parameters for distorted pulses), forward propagation computation (extracting features of the pulse signals and identifying their shaping parameters through the transformer model), and backward propagation computation (optimizing model parameters using error backpropagation to improve identification accuracy).

3.2 Dataset

Taking the third-order pulse as an example, the dataset used for the Transformer model in this paper is composed of the sampled values of the trapezoidal pulse obtained after trapezoidal shaping of the distorted pulse, along with its corresponding (\times 个) set of trapezoidal shaping parameters. Let denote the set of trapezoidal shaping parameters, which is derived from the relevant parameters of the distorted pulse and the timing parameters associated with its trapezoidal

shaping. Taking the i th parameter set P_i as an example, this set includes the pulse amplitude (where $i = 1, 2, \dots, N$), time constants, as well as the rise time and flat-top duration of the trapezoidal shaping. The dataset can be represented in matrix form as follows:

The dataset (25) comprises N samples of distorted pulses, with each row representing the data of a sample. The initial m columns of data in each sample correspond to the sampled values of the corresponding trapezoidal shaping pulse for the sample, and the final column represents the set of trapezoidal shaping parameters. The parameters expressed as $[1, 2, \dots, \dots]$ are all controlled to be within a small range when creating the dataset. It should be noted that, as previously mentioned, the dataset is divided into a training set and a validation set with a ratio of 8:2. The training set is used to train the model and determine parameters such as weights and biases. The validation set, on the other hand, is not involved in the training process and is only used to adjust hyperparameters of the model, such as the number of training epochs and the learning rate. A sufficient test set is also prepared under the same conditions to evaluate the model's performance after training. When creating the dataset, the trapezoidal shaping results of three types of distorted pulses (first-, second-, and third-order) are shown in Figure 6 [Figure 6: see original paper]. Before the pulses become distorted, the amplitudes of the trapezoidal shaping results for all three types of pulses are 150 mV. After distortion, a common feature can be observed in the three trapezoidal pulses: a significant loss in the amplitude of the shaping results.

3.3 Forward Propagation

During the forward propagation of the model, the sampled values of the trapezoidal pulses from the training set, denoted as X , are processed through positional encoding and then fed into the encoder as input. The encoder calculates global \times dependencies through the attention mechanism, extracts sequence features, and sequentially passes them to the normalization residual connection module and the feedforward network. Ultimately, the output is generated through a linear layer: the set of trapezoidal shaping parameters θ . Unlike traditional RNNs, the Transformer processes sequences based on the attention mechanism without relying on recursive relationships between time steps. Since the Transformer processes the sequence as a whole, it can fully leverage parallel computing, significantly reducing training time and enhancing its ability to model long sequences. In addition, positional encoding is a crucial part of the input stage. It provides the model with sequence order information through sine and cosine functions, thereby compensating for the Transformer's insensitivity to sequence order.

(1) Positional Encoding: In the Transformer architecture, since recurrent neural networks or convolutional neural networks are not used, positional information within the sequence cannot be directly utilized. Therefore, to enable the model to learn information about various positions in the sequence, the Transformer employs positional encoding. The positional encoding is calculated as

follows:

Let the input to the model be X . Then X is obtained by adding the trapezoidal pulse sequence and the positional encoding PE of the pulse sequence. That is, after positional encoding, the expression for the input sequence X to the model is as follows:

(2) Self-Attention Mechanism: The core of the Transformer model is the self-attention mechanism. This mechanism calculates the relevance weights of each element with respect to other elements in the sequence, and then uses these weights to compute the weighted sum for each element. The use of the self-attention mechanism not only increases computational efficiency, but also allows for a better capture of the correlations between elements.

Given the sequence X obtained after positional encoding of the trapezoidal pulse sampled values, in the self-attention mechanism, for each input sequence X , three different linear transformations are first applied to compute the query vector Q , the key vector K , and the value vector V . This linear transformation process can be expressed as:

Here W_Q , W_K , W_V , W_O are the weight matrices for model training. Then, the attention weights are calculated by dot product, and the final output of the self-attention mechanism weights is shown in Equation (31): where d_k is the dimension of the vector K , and Q , K and V represent the data vectors of the input sequence X mapped to query, key and value respectively.

(3) Multi-Head Attention: The multi-head attention mechanism can be understood as a mechanism that maps the original sets of Q (query), K (key), and V (value) vectors into h subspaces and independently computes attention weights in each subspace. Specifically, the Q , K , and V vectors, obtained from the input sequence X through Equation (31), are linearly transformed into vectors in h subspaces, respectively. Then, dot-product attention weights are computed separately in each subspace. Finally, the attention information obtained from each subspace is concatenated and passed through another linear transformation to yield the final output vector of the multi-head attention layer. The multi-head attention mechanism is calculated as follows:

Here, W_{Q_i} , W_{K_i} , W_{V_i} represent the linear projection parameter matrices for the i -th head, W_O represents the parameter matrix for the output projection, and h_i denotes the self-attention distribution of the i -th head.

(4) Feed-Forward Neural Network: In the Transformer model, the Feed-Forward Network (FFN) receives the output from the Multi-Head Attention (MHA) mechanism and Layer Normalization as its input. It then maps this input to a new representation vector through two linear transformations followed by a nonlinear activation function. Assuming Y is the output obtained from the input sequence X after processing through the multi-head attention layer and the normalized residual connection, the computation process of the feed-forward network can be expressed as:

where W_1 and W_2 represent two different sets of weights, and b_1 and b_2 are the two bias parameters.

(5) Normalization and Residual Connections: The data passing through the multi-head attention or feed-forward network in a Transformer model is not directly used as the output. Instead, it undergoes normalization and residual connections for further processing. The primary purpose of normalization is to prevent gradient vanishing and gradient explosion, thereby accelerating training speed and enhancing training stability. The addition of residual connections helps address the degradation problem in neural networks. Let x be the output obtained after processing the input sequence X through the feed-forward neural network layer and the normalization with residual connections. The mathematical model for normalization and residual connections can be expressed as:

Here, y represents the result after multi-head attention and normalization; $MHA(X)$ and $FFN(\cdot)$ denote the outputs of the multi-head attention and feed-forward network, respectively.

(6) Linear Layer: Finally, a linear layer is applied to map the high dimensional features from the Transformer encoder to the values of the pulse trapezoidal shaping parameters, resulting in the trapezoidal shaping parameter set P . This output layer is implemented using a linear transformation and its mathematical expression is as follows:

where W represents the weight matrix, b is the bias vector, x is the output vector from the Transformer encoder, and P is the final predicted set of pulse parameters. In summary, the forward propagation calculation process is shown in Figure 7 [Figure 7: see original paper].

3.4 Backpropagation

In the Transformer model, the initial values of the three matrices W_1 , W_2 , and W_3 corresponding to the input sequence X are randomly initialized. During each forward propagation, the model's output (i.e., the set of predicted trapezoidal shaping parameters P) is computed and compared to the actual set of trapezoidal shaping parameters from the training set to obtain the loss value (or error). Therefore, during the backpropagation process, the error is calculated using the loss function, and the gradients of the model parameters W_1 , W_2 , and W_3 are updated based on this error. This allows the model to gradually reduce the error and improve prediction accuracy in subsequent iterations. For a training set comprising M samples, the Mean Absolute Error (MAE) of the parameter set is used as the value of the loss function, which is calculated as follows:

In summary, the network architecture of the Transformer, as illustrated in Figure 8 [Figure 8: see original paper], includes the input layer, various encoder sublayers, the output layer, and the backpropagation structure.

3.5 Model Training

As mentioned above, in order to investigate the effectiveness of the combined method of Frequency Domain Analysis (hereafter abbreviated as FDA) and the Transformer model in identifying parameters related to distorted pulses, this study performed classification training on the three types of distorted pulses mentioned above (first, second and third order pulses) along with their initial parameter value ranges. This resulted in three transformer models with different parameter configurations, named Transformer1, Transformer2 and Transformer3, each consisting of several sub-models. These three models were trained using the Adam optimizer, and the key hyperparameter configurations for some of the models are shown in Table 1 .

After training, the loss function iterations of the training and validation sets for each of the three sub-models taken from the three transformer models are shown in Figure 9 [Figure 9: see original paper]: Figure 9.a shows the values for Transformer1, Figure 9.b shows the loss values for Transformer2, and Figure 9.c shows the loss values for Transformer3. As can be seen in Figure 9, as the number of training cycles increases, the loss values of the training and validation sets of the three models show a downward trend, with some minor fluctuations in between, and eventually tend to stabilize. This indicates that the model is not overfitting and therefore does not require a separate dropout layer.

3.6.1 Restoration of First-Order Distortion Pulses

Taking the two distorted output pulses from Eq. (14) as an example, they are denoted as pulse 1 and pulse 2. It is assumed that the amplitude coefficient A of the pulse to be restored is 300 mV, with time constants τ of 3.6 s and 7.4 s respectively. At the same time, two sub-models, denoted Transformer1' and Transformer1'', are extracted from the first-layer model. These sub-models are trained using time constant ranges of [3.2, 5.0] s and [6.0, 10] s, respectively. The trapezoidal shaping results of distorted Pulse 1 and Pulse 2 are then fed into these two sub-models for analysis. Figure 10 [Figure 10: see original paper] shows a comparison between the two original first order distorted pulses and the pulses identified and restored by each model. Table 2 shows a comparison of the true pulse amplitude parameters and the pulse amplitude parameters identified by each sub-model, both before and after the incorporation of frequency domain analysis (FDA). The comparison shows that the average relative error of the pulse amplitude coefficients is 1.02% before the application of FDA, whereas it is reduced to 0.21% after the application of FDA. This shows that FDA can accurately select the model that corresponds to the range of pulse parameter values, thereby allowing more accurate identification of the pulse parameters.

3.6.2 Restoration of Second-Order Distortion Pulses

In the experiment described in Section 3.6.1, we analyzed first-order distorted pulses and verified the feasibility of combining frequency domain analysis with

the Transformer model. Now, we will conduct an experiment using two second-order distorted pulses generated by Eq. (16). The relevant parameters of the two pulses are presented in Table 3. Simultaneously, two sub-models are selected from the second-layer model: the Transformer2' model, trained with time constant ranges of $3.2 \leq \tau_1 \leq 5.0$ and $1.2 \leq \tau_2 \leq 2.8$; and the Transformer2'' model, trained with time constant ranges of $5.2 \leq \tau_1 \leq 8.0$ and $2.0 \leq \tau_2 \leq 2.8$; the unit for these time constants is μs . The trapezoidal shaping results of distorted Pulse 1 and Pulse 2 are then fed into these two sub-models for analysis. Figure 11 [Figure 11: see original paper] presents a comparison between the original second-order distorted pulses and the pulses identified and restored by each model. Table 4 compares the true amplitude parameters of the pulse before and after the addition of frequency domain analysis with the pulse amplitude parameters identified by each sub-model. Before the addition of the frequency domain analysis (FDA), the average relative error of the pulse amplitude coefficient was 1.22%; after the addition of the frequency domain analysis (FDA), the average relative error of the pulse amplitude coefficient was 0.35%.

3.6.3 Restoration of Third-Order Distortion Pulses

For the restoration of distorted third-order pulses, more parameters need to be identified, placing higher demands on the model. Directly identifying pulses over a wide range would increase the difficulty of identification and make it hard to achieve satisfactory results. Therefore, the introduction of frequency domain analysis significantly narrows the pulse search range and enables the model to achieve better identification performance within each smaller range. To further verify the feasibility of combining frequency domain analysis with Transformer for distorted third-order pulses, two such pulses are selected based on Equation (13) and labeled as Pulse 1 and Pulse 2. The relevant parameters of Pulse 1 and Pulse 2 are shown in Table 5. Two sub-models are selected from the third-layer model: the Transformer3' model, trained with time constant ranges of $1 \leq \tau_1 \leq 2.8$, $12 \leq \tau_2 \leq 14$, and $50 \leq \tau_3 \leq 52$; and the Transformer3'' model, trained with time constant ranges of $1 \leq \tau_1 \leq 2.8$, $8 \leq \tau_2 \leq 10$, and $60 \leq \tau_3 \leq 62$; the unit for these time constants is μs . The trapezoidal shaping results of distorted Pulse 1 and Pulse 2 are then fed into these two sub-models for analysis. Figure 12 [Figure 12: see original paper] presents a comparison between the original third-order distorted pulses and the pulses identified and restored by each model. Table 6 shows a comparison of the true pulse amplitude parameters and the pulse amplitude parameters identified by each sub-model, before and after incorporating frequency domain analysis (FDA). Before adding FDA, the average relative error of the pulse amplitude coefficients was 1.23%; after incorporating FDA, the average relative error of the pulse amplitude coefficients was reduced to 0.46%. It is evident that the addition of FDA has improved the model's performance, enhancing its parameter identification capabilities compared to before.

3.7.1 Qualitative Analysis

To comprehensively and deeply evaluate the effectiveness of the combined method of frequency domain analysis (FDA) and Transformer model in distorted pulse correction, as well as its optimization and improvement for energy spectrum analysis, copper ore samples were selected as the research object in this experiment. The measurement system included a high-performance Silicon Drift Detector (FAST-SDD) and an X-ray tube. The detector was a FAST-SDD detector produced by Amptek, with the model number FAST-RC101. The data acquisition board, featuring a 20MHz 14bit ADC, was developed by Xinxianda Measurement and Control Technology Co., Ltd. The current of the X-ray tube could be adjusted within the range of 3.9uA to 1000uA, and the voltage could be adjusted within the range of 0 to 50kV. Through the analysis of the measurement results, we obtained the X-ray spectrum, as shown in Figure 13 [Figure 13: see original paper]. The black spectrum in the figure reveals the presence of unknown shadow peaks preceding the characteristic peaks with higher count rates, in the absence of any correction. It is speculated that these shadow peaks are caused by the phenomenon known as count drift. Count drift refers to the loss of amplitude that occurs during the trapezoidal shaping of distorted pulses, leading to the transfer of counts originally assigned to the characteristic peak region to the shadow peak region. This results in the formation of prominent shadow peaks on the left side of the characteristic peaks.

The existence of these shadow peaks poses a significant risk of being mistaken for characteristic peaks of certain elements, thereby directly affecting the accurate analysis of element content. Therefore, it is crucial to correct the shadow peaks in the energy spectrum. During the correction process, we first processed the trapezoidal shaping results of the distorted pulses using a pre-trained model. Then, we replaced the amplitudes of the original distorted pulses with the amplitudes estimated by the model for the trapezoidal pulses. After this series of processing steps, we obtained optimized results, as indicated by the red spectrum in Figure 13, where the shadow peaks were effectively corrected. Figure 13 presents a comparison of the spectrum before and after the correction of the shadow peaks, demonstrating the efficacy of the proposed method that combines frequency domain analysis with the Transformer model in correcting shadow peaks induced by pulse distortion.

3.7.2 Quantitative Analysis

As mentioned earlier, the loss of amplitude in distorted pulses leads to count drift, resulting in the formation of a shadow peak on the left side of the characteristic peak. To quantitatively analyze the restoration effect of the energy spectrum counts, we use the peak area (Area) to represent the total count within a specific channel range. Therefore, the analysis of the energy spectrum count restoration results can be transformed into an analysis of the restoration ratio of the peak area. The following is the formula for calculating the peak area:

In the equation: denotes the count of the restored portion of the shadow peak after implementing spectral peak correction using the model, as depicted in Figure 14a [Figure 14: see original paper]. It is numerically equivalent to the difference in peak area within the channel address interval 92-140, where the shadow peak resides, before and after the spectral peak correction. represents the count of the restored portion of the characteristic peak after implementing spectral peak correction using the model, as illustrated in Figure 14b. It is numerically equal to the difference in peak area within the channel address interval 162-212, where the characteristic peak is located, before and after the spectral peak correction.

In order to better analyze the model, an index R is introduced to quantify the model's performance. R represents the proportion of shadow peaks that the model corrects, which is numerically equal to the ratio of the difference in peak areas between the address intervals where the feature peaks are located before and after the model is applied to the peak area of the shadow peak generated by the distortion pulse. The calculation is as follows:

Due to the randomness of radioactive measurements and the inevitable statistical errors in the measurement process, there may be some variation in the correction ratio of the shadow peak. Therefore, in this paper, five experiments were performed on the sample and the resulting discrete trapezoidal pulse sequences were used as input to the model. The pulse amplitudes identified by the model were then used to replace the original pulse amplitudes and the results were analyzed using formulae (39), (40) and (41) as shown in Table 7.

From the measurement results presented in Table 7, it is evident that the corrected spectra obtained through pulse amplitude recognition using the Transformer model exhibit two notable characteristics. Firstly, the area of the characteristic peaks has increased compared to the original spectra. Secondly, there is a significant reduction in the area of the shadow peaks. According to the principle of count drift, the area of the shadow peaks should be close to the area increment of the characteristic peaks restored after pulse parameter identification by the model. The effectiveness of the model's correction is evaluated using the R -indicator, and as shown in the table, approximately 94.15% of the shadow peak area can be corrected into characteristic peaks. This holds great significance for improving the accuracy of radioactive measurements.

4 Conclusion

To solve the problem of complex and distorted output signals from the front-end analogue system in nuclear pulse signal detection, this paper proposes a pulse distortion correction method that combines frequency domain analysis with the transformer model. By using frequency domain analysis, the property and initial values of the components contained in the pulse signal are obtained. Then, based on the characteristics of the pulse components and initial parameters, different transformer models are trained to identify the relevant parameters of

the distorted pulses.

Simulations were performed to compare the relative error of the models in identifying the pulse amplitudes before and after incorporating the frequency domain analysis. The results indicate that the relative error of the model is reduced by approximately 0.8% after the integration of the frequency domain analysis. This result verifies the effectiveness and accuracy of the proposed method in correcting distorted pulses.

In the experiment, we performed five X-ray measurements on a copper ore sample using FAST-SDD and processed the pulses using trapezoidal shaping and multi-channel pulse amplitude analysis. Finally, an X-ray energy spectrum corrected for shadow peaks was obtained. The experimental results show that the proposed model successfully identifies the amplitude parameters of the measured pulse sequences. Meanwhile, using the original energy spectrum obtained without invoking the model as a reference, a comparison with the corrected energy spectrum shows that approximately 94.15% of the shadow peaks can be corrected to characteristic peaks. This is of great importance for improving the completeness and resolution of the X-ray energy spectrum count.

In summary, the distortion pulse restore method proposed in this paper, which is based on frequency domain analysis and the Transformer model, offers a novel approach to addressing nuclear pulse distortion issues. Furthermore, this method demonstrates significant potential in improving the accuracy and resolution of energy spectrum counting, highlighting its important application prospects.

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